

## 1. Features

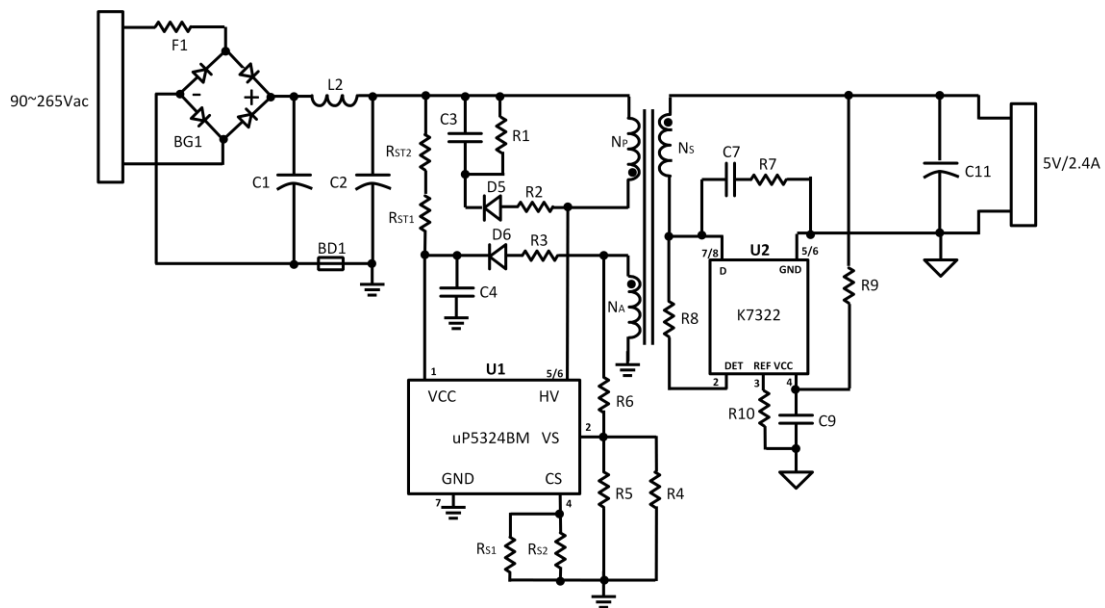
- *Built-in 18mΩ on-resistance and 45V breakdown voltage MOSFET*
- *Programmable resistor to filter resonant ringing signals*
- *Fewest components counts*

- *Excellent system ESD and EFT performance*

## 2. Applications

- *Smart phone chargers*
- *Power strip with USB ports*
- *5V adapters*

### 3. Typical applications (5V/2.4A charger)

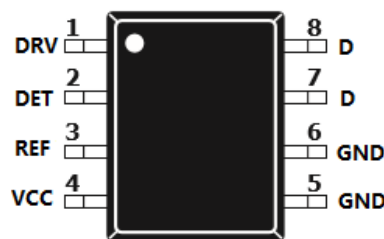


*Fig.1 Typical applications*

## 4. Ordering information

<i>Part number</i>	<i>Mark ID</i>	<i>Package</i>	<i>Quantity per reel</i>	<i>Output power</i>
<i>K7322M-45</i>	<i>7322M-45</i>	<i>SOP-8</i>	<i>3,000</i>	<i>10~12W</i>

## 5. Pin definitions



*K7322M-45*

<i>Pin Name</i>	<i>Pin Type</i>	<i>Pinout</i>	<i>Pin Functions</i>
<i>DRV</i>	<i>Output</i>	<i>1</i>	<i>Gate drive of SR MOSFET</i>
<i>DET</i>	<i>Input</i>	<i>2</i>	<i>Drain voltage detection of SR MOSFET</i>
<i>REF</i>	<i>Output</i>	<i>3</i>	<i>Volt-second setting for SR MOSFET turn on</i>
<i>VCC</i>	<i>Power supply</i>	<i>4</i>	<i>Power supply of the rectifier</i>

GND	Ground	5, 6	Source of the SR MOSFET and the power ground.
D	Drain	7, 8	Drain of the SR MOSFET

## 5. Absolute maximum ratings (Note 1)

Parameter	Name	Range	Unit
Voltage range	VCC	-0.3 to 7	V
Voltage range	DRV	-0.3 to 6	V
Voltage range	DET	-2 to 50	V
Voltage range	REF	-0.3 to 6	V
Voltage range	D	-2 to 45	V
Continuous drain current	$I_D$	10	A
Pulse drain current	$I_{DP}$	40	A
Power dissipation @ $T_A=25^\circ\text{C}$	$P_D$	0.8	W
Maximum junction temperature	$T_{JMAX}$	150	$^\circ\text{C}$
Lead temperature	$T_{LEAD}$	300	$^\circ\text{C}$
Storage temperature	$T_{STG}$	-55 to 150	$^\circ\text{C}$
ESD rating per JEDEC JESD22-A114	HBM	2000	V
ESD rating per JEDEC JESD22-C101C	CDM	1000	V
Latchup test per JEDEC 78		+/-200	mA

Note1: Stresses over those listed under “Absolute maximum ratings” may cause permanent damages to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “Recommended Operating Conditions” is not implied. Exposure to “Absolute Maximum Ratings” for extended periods of time may affect device reliability.

## 6. Thermal parameter

Junction to ambient thermal resistance	$\theta_{JA}(SOP-8)$	90	$^\circ\text{C/W}$
Junction to case thermal resistance	$\theta_{JC}(SOP-8)$	45	$^\circ\text{C/W}$

## 7. Recommended operating conditions

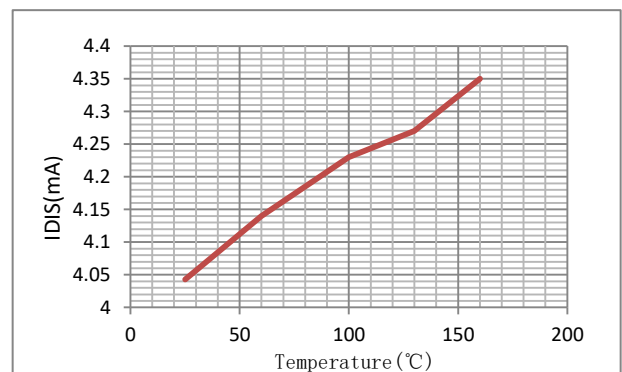
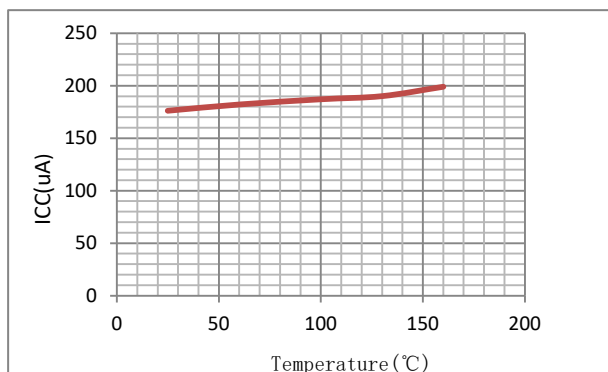
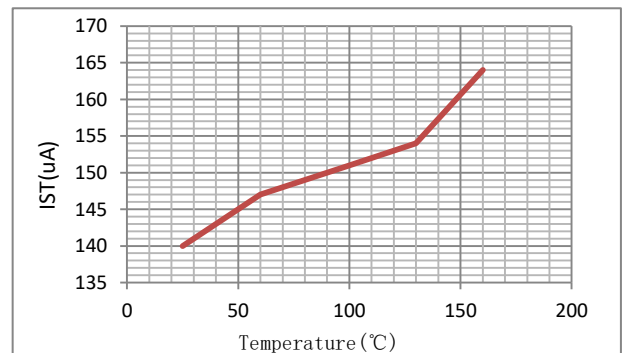
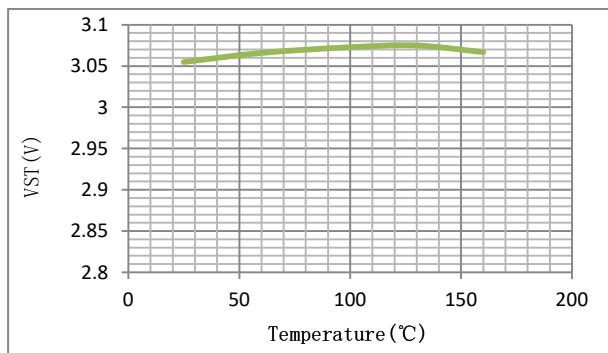
Parameter	Symbol	Min	Max	Unit
Supply voltage	VCC	3.3	6	V
Ambient Temperature	$T_A$	-40	85	$^\circ\text{C}$

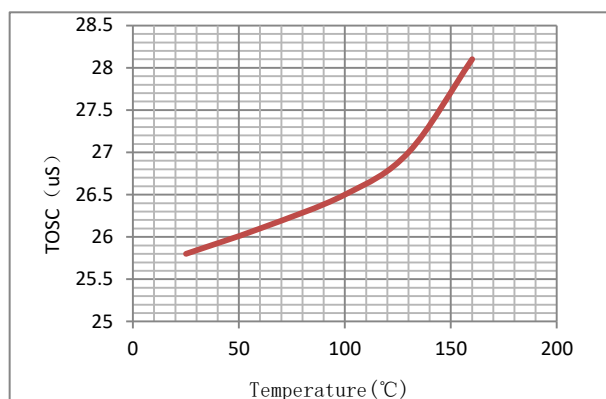
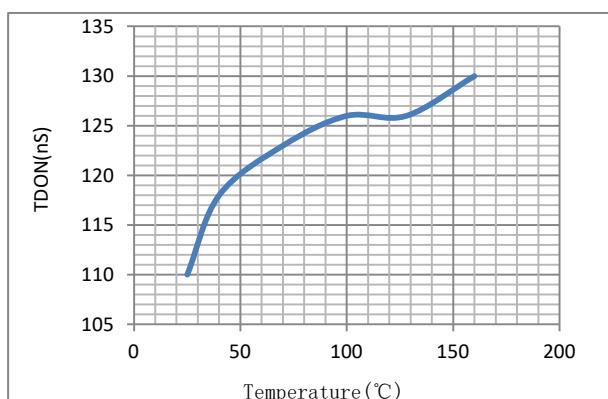
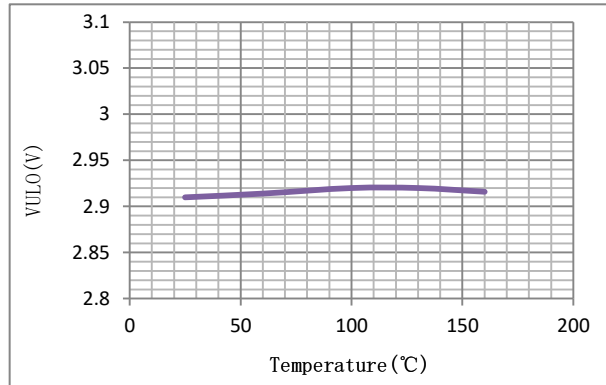
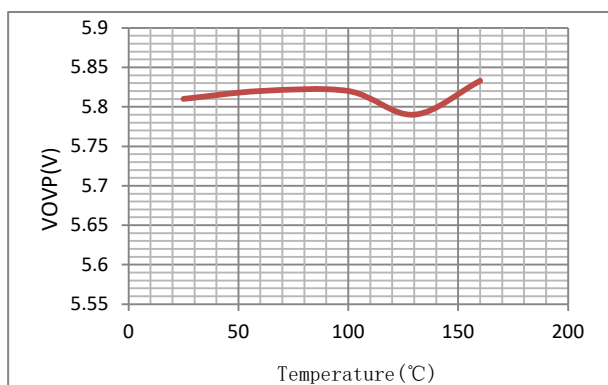
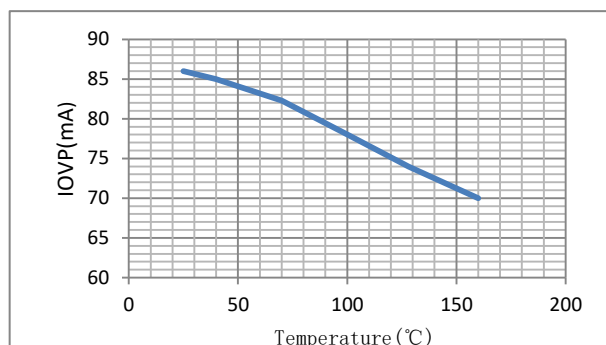
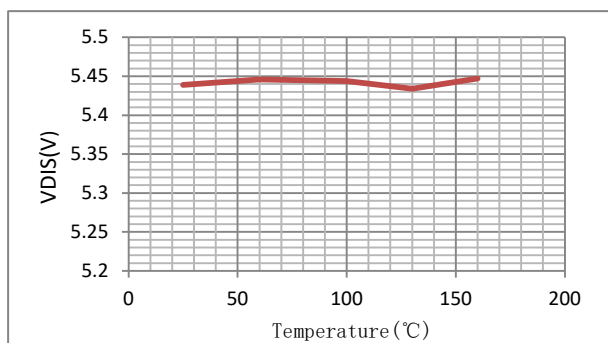
## 8. Electrical parameters

$T_A = 25^\circ\text{C}$ , unless otherwise specified

Parameter	Symbol	Condition	Min	Typ	Max	Unit
<b>Power supply(VCC pin)</b>						
Operating current	$I_{CC}$		180	200	220	$\mu\text{A}$
Startup voltage	$V_{ST}$		2.68	2.98	3.28	V
Minimum operating voltage	$V_{UVLO}$		2.5	2.8	3.1	V
Startup current	$I_{ST}$	$V_{CC} = V_{ST} - 0.1V$	110	140	170	$\mu\text{A}$
<b>Output voltage monitor</b>						
VCC discharge voltage	$V_{DIS}$		5.39	5.47	5.55	V
VCC discharge current	$I_{DIS}$	$V_{CC} = V_{DIS} + 0.1V$	1.5	3	4.5	mA

VCC protection voltage	$V_{OVP}$		5.75	5.85	5.95	V
VCC over voltage discharge current	$I_{OVP}$	$V_{CC} = V_{OVP} + 0.1V$	40	97	130	mA
VCC OVP discharge time	$T_{OVPDIS}$		672	800	928	$\mu S$
Internal oscillator period	$T_{OSC}$		21	25	29	$\mu S$
<b>Synchronous rectification control</b>						
SR turn on voltage	$V_{THON}$			75		mV
SR turn off voltage	$V_{THOFF}$		-6	-2	2	mV
SR turn on delay time	$T_{DON}$		10	70	130	nS
SR turn off delay time	$T_{DOFF}$		10	100	150	nS
SR turn on rising time	$T_R$	$C_L = 4.7nF$	10		100	nS
SR turn off falling time	$T_F$	$C_L = 4.7nF$	10		100	nS
SR minimum on time	$T_{LEB\_S}$	$(V_{DET} - V_{CC}) * T_{ONP} = 30V * \mu S$		2.2		$\mu S$
SR minimum operating voltage( $V_{DET} - V_{CC}$ )	$V_{S\_MIN}$	Minimum DET pin voltage@ $V_{CC} = 5V$		3.0		V
Ampere Second Product	ASP	$(V_{DET} - V_{CC}) * T_{ONP} = 25V * \mu S$	0.5	0.7	0.9	mA* $\mu S$
<b>SR MOSFET characteristics</b>						
Drain to source breakdown	$BV_{dss}$	$V_{GS} = 0V, I_D = 0.25mA$	45			V
Gate threshold voltage	$V_{GS(TH)}$	$V_{DS} = V_{GS}, I_D = 0.25mA$	1.0	1.5	2.0	V
Static Drain-to-Source On Resistance	$R_{DS(on)}$	$V_{GS} = 5.5V, I_D = 15A$		18	25	$m\Omega$
Drain-to-Source leakage	$I_{DSS}$	$V_{GS} = 0V, V_{DS} = 50V$			1	$\mu A$
Gate to source leakage	$I_{GSS}$	$V_{GS} = \pm 20V$	-100		100	nA





## 9. Functional block diagram

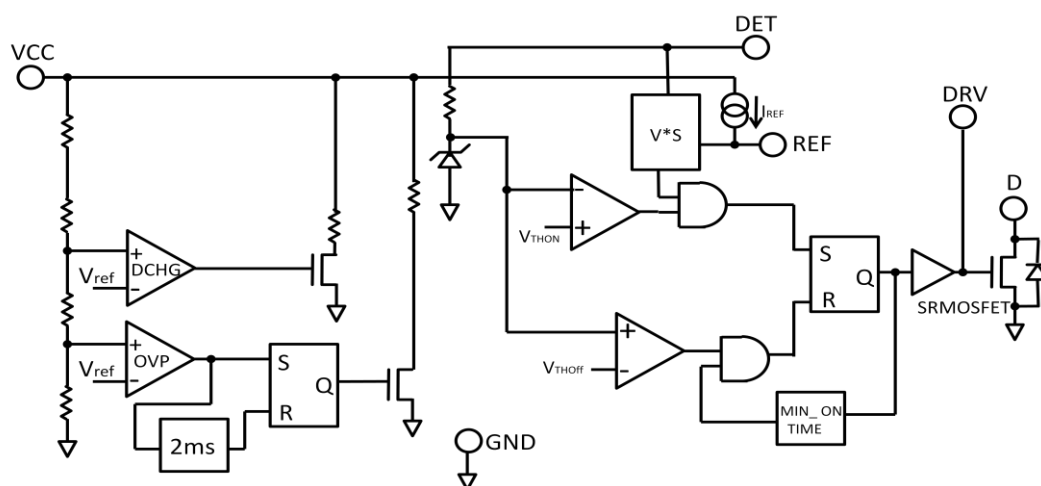


Fig.2, K7322 block diagram

## 10. Principle of operation

The K7322 has two major functions: output voltage monitor to discharge output capacitor at certain conditions and synchronous rectification. The device must work in Discontinuous Conduction Mode (DCM) or Quasi-Resonant Mode (QRM).

### 10.1 Power up and power down sequences

Refer to Fig.1 and Fig.2, after AC power supply is applied to the converter, the primary controller UP5324A (U1) starts to deliver energy to the output capacitor C11, the output voltage begins rising from 0V. When the VCC voltage of K7322 (U2) is lower than the startup voltage  $V_{ST}$ , the synchronous rectifier does not work, the body diode of the SR MOSFET acts as the rectification diode, with around -1.5V forward conduction voltage since the body diode of the SR MOSFET is just an ordinary PN junction. When the VCC voltage of K7322 (U2) is larger than the startup voltage  $V_{ST}$ , the synchronous rectifier starts to work, as described in 10.3. When the AC power supply is removed from the converter, the VCC voltage of K7322 (U2) falls below  $V_{VLO}$ , the synchronous rectifier stops working, the body diode of the SR MOSFET acts again as the rectification diode.

### 10.2 Discharge of output capacitor

When the VCC voltage of K7322 (U2) is higher than a specified voltage  $V_{DIS}$ , K7322 will turn on a discharge path from VCC to GND with typical 3mA current capacity to make the system output voltage stay around  $V_{DIS}$ . When the VCC voltage of K7322 (U2) is further higher than a specified voltage  $V_{OVP}$ , such as in case of load transient from full load to no load, K7322 will turn on another discharge path from VCC to GND with typical 70mA current capacity to limit the system output over shoot voltage.

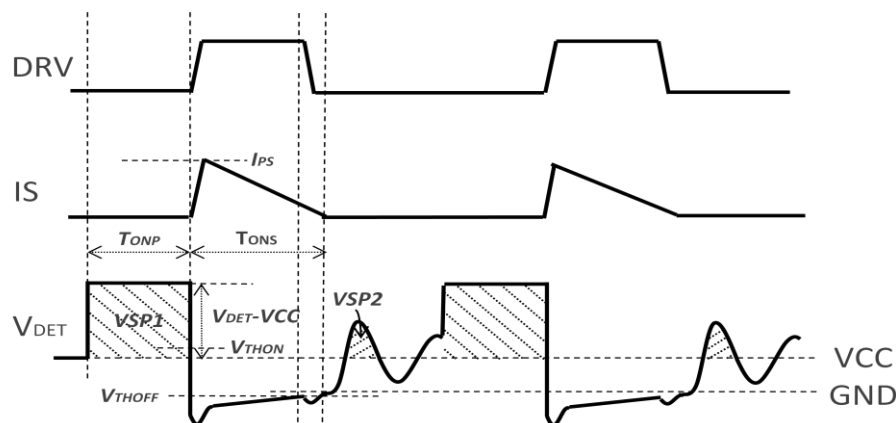


Fig.3, K7322 SR operation

### 10.3 Synchronous rectification

Refer to Fig3, K7322 monitors the SR MOSFET drain to source voltage at DET pin. When the  $V_{DET}$  is lower than the turn-on threshold voltage  $V_{THON}$ , K7322 DRV pin generates a positive drive voltage after a turn-on delay time ( $T_{DON}$ ). The SR MOSFET will turn on and the current will transfer from the body diode to the channel of the SR MOSFET.

After the conduction of the SR MOSFET, the  $V_{DET}$  rises linearly. When it rises over the turn off threshold voltage  $V_{THOFF}$ , K7322 DRV pin generates a pull down signal after a turn-off delay ( $T_{DOFF}$ ).

During the SR MOSFET turn on process, some ringing noise may be generated. The minimum on-time block blanks the output of  $V_{THOFF}$  comparator, keeping the SR MOSFET on for at least the

minimum on time. The minimum on time is proportional to the volt second product (VSP) of the primary side power switch on state, which is equal to  $(V_{DET}-V_{CC}) \cdot T_{ONP}$ . If  $(V_{DET}-V_{CC}) \cdot T_{ONP} = 30V \cdot \mu S$ , the minimum on time is about  $2.2\mu S$ .

As the convertor operates in DCM or QRM, after synchronous rectifier stops conduction, resonant ringing is resulted due to the primary inductance and power switch parasitic capacitance. This ringing waveform may leads to the error conduction of the synchronous rectifier. To avoid this fault, K7322 judges the primary power switch turn on by the Volt-Second Product (VSP) of the system. The volt-second product (VSP1) of a primary switch turn on is much higher than the volt-second product (VSP2) of the resonant ringing waveform, as illustrated in Fig.3. Thus, before to turn on the synchronous rectifier, K7322 judges if the detected volt-second product of  $V_{DET}$  voltage above  $V_{CC}$  is higher than a threshold ( $VSP_{REF}$ ) and then turn on synchronous rectifier if the detected VSP is larger than  $VSP_{REF}$ . The purpose of REF resistor is to set the volt-second product threshold ( $VSP_{REF}$ ). The detected volt-second product

$$VSP = \int (V_{DET}-V_{CC}) \cdot dt = R_{VSP} \cdot C_{VSP} \cdot V_{VSP}$$

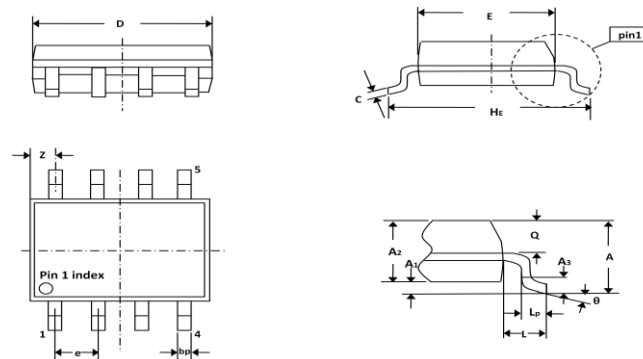
Where  $R_{VSP}$  is an internal resistor to convert the  $(V_{DET}-V_{CC})$  voltage to charge an internal capacitor  $C_{VSP}$ , and  $V_{VSP}$  is the voltage rise on  $C_{VSP}$  when the charging phase is completed. In K7322, the  $V_{VSP}$  to judge whether the VSP is generated by primary side power switch or not is set by an internal current source  $I_{REF}$  and outside resistor  $R_{REF}$  ( $R10$  in Fig.1). So, the judging volt-second product

$$VSP_{REF} = (R_{VSP} \cdot C_{VSP} \cdot I_{REF}) \cdot R_{REF} = ASP \cdot R_{REF}$$

Where  $ASP = (R_{VSP} \cdot C_{VSP} \cdot I_{REF})$  is a K7322 determined "Ampere Second Product" to set system  $VSP_{REF}$  with  $R_{REF}$ .  $VSP_{REF}$  depends on system design and are always fixed after system design is frozen.  $R_{REF}$  resistor should be considered for the worst case, that is, the minimum primary peak current condition.  $VSP_{REF}$  should be designed in the middle of VSP1 and VSP2. K7322 also sets a minimum line voltage to operate the SR MOSFET. The value of  $V_{DET}-V_{CC}$  during primary side power switch turn on time ( $T_{ONP}$ ) must be higher than  $V_{S\_MIN}$  to enable the synchronous rectifier. That is, the minimum rectified input line voltage ( $V_{IN\_MIN}$ ) to enable the SR is

$$V_{IN\_MIN} = V_{S\_MIN} \cdot (N_P/N_S)$$

## 11. Mechanical dimensions



UNIT	A	A1	A2	A3	bp	c	D	E	e	He	L	Lp	Q	θ
mm	1.75	0.1/ 0.25	1.25/ 1.5	0.25	0.33/ 0.51	0.19/ 0.25	4.7/ 5.1	3.8/ 4.0	1.27	5.8/ 6.2	1.05	0.4/ 1.0	0.6/ 0.7	8°